

LF147/LF347

Wide Bandwidth Quad JFET Input Operational Amplifiers

General Description

The LF147 is a low cost, high speed quad JFET input operational amplifier with an internally trimmed input offset voltage (BI-FET II™ technology). The device requires a low supply current and yet maintains a large gain bandwidth product and a fast slew rate. In addition, well matched high voltage JFET input devices provide very low input bias and offset currents. The LF147 is pin compatible with the standard LM148. This feature allows designers to immediately upgrade the overall performance of existing LF148 and LM124 designs.

The LF147 may be used in applications such as high speed integrators, fast D/A converters, sample-and-hold circuits and many other circuits requiring low input offset voltage, low input bias current, high input impedance, high slew rate and wide bandwidth. The device has low noise and offset voltage drift.

Features

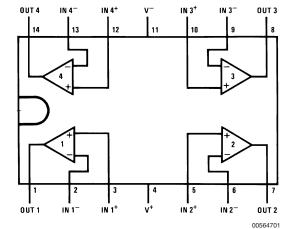
Internally trimmed offset voltage:Low input bias current:	5 mV max 50 pA
■ Low input noise current:	0.01 pA/√Hz
Wide gain bandwidth:High slew rate:	4 MHz 13 V/μs
Low supply current:High input impedance:	7.2 mA 10 ¹² Ω
■ Low total harmonic distortion:	≤0.02%
Low 1/f noise corner:Fast settling time to 0.01%:	50 Hz 2 μs

Simplified Schematic

VCC O VO VO INTERNALLY TRIMMED TRIMMED O0564713

Connection Diagram

Dual-In-Line Package



Note 1: LF147 available as per JM38510/11906.

Top View

Order Number LF147J, LF147J-SMD, LF347M, LF347BN, LF347N, LF147J/883, or JL147 BCA (Note 1)
See NS Package Number J14A, M14A or N14A

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Absolute Maximum Ratings (Note 2)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

	LF147	LF347B/LF347
Supply Voltage	±22V	±18V
Differential Input Voltage	±38V	±30V
Input Voltage Range	±19V	±15V
(Note 3)		
Output Short Circuit	Continuous	Continuous
Duration (Note 4)		
Power Dissipation	900 mW	1000 mW
(Notes 5, 11)		
T _j max	150°C	150°C
$ heta_{jA}$		
Ceramic DIP (J) Package		70°C/W
Plastic DIP (N) Package		75°C/W
Surface Mount Narrow (M)		100°C/W
Surface Mount Wide (WM)		85°C/W

	LF147	LF347B/LF347
Operating Temperature	(Note 6)	(Note 6)

Range

Storage Temperature

Range $-65^{\circ}\text{C} \le \text{T}_{A} \le 150^{\circ}\text{C}$

Lead Temperature

(Soldering, 10 sec.) 260°C 260°C

Soldering Information

Dual-In-Line Package

Soldering (10 seconds) 260°C

Small Outline Package

Vapor Phase (60 seconds) 215°C Infrared (15 seconds) 220°C

See AN-450 "Surface Mounting Methods and Their Effect on Product Reliability" for other methods of soldering

surface mount devices.

ESD Tolerance (Note 12) 900V

DC Electrical Characteristics (Note 7)

Symbol	Parameter	Conditions	LF147		LF347B			LF347			Units	
			Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	
V _{os}	Input Offset Voltage	R _S =10 kΩ, T _A =25°C		1	5		3	5		5	10	mV
		Over Temperature			8			7			13	mV
$\Delta V_{OS}/\Delta T$	Average TC of Input Offset	R _S =10 kΩ		10			10			10		μV/°C
	Voltage											
I _{os}	Input Offset Current	T _j =25°C, (Notes 7, 8)		25	100		25	100		25	100	pА
		Over Temperature			25			4			4	nA
I _B	Input Bias Current	T _j =25°C, (Notes 7, 8)		50	200		50	200		50	200	pА
		Over Temperature			50			8			8	nA
R _{IN}	Input Resistance	T _j =25°C		10 ¹²			10 ¹²			10 ¹²		Ω
A _{VOL}	Large Signal Voltage Gain	V _S =±15V, T _A =25°C	50	100		50	100		25	100		V/mV
		$V_O=\pm 10V$, $R_L=2 k\Omega$										
		Over Temperature	25			25			15			V/mV
Vo	Output Voltage Swing	$V_S=\pm 15V$, $R_L=10 \text{ k}\Omega$	±12	±13.5		±12	±13.5		±12	±13.5		V
V _{CM}	Input Common-Mode Voltage	V _S =±15V	±11	+15		±11	+15		±11	+15		V
	Range			-12			-12			-12		V
CMRR	Common-Mode Rejection Ratio	R _S ≤10 kΩ	80	100		80	100		70	100		dB
PSRR	Supply Voltage Rejection Ratio	(Note 9)	80	100		80	100		70	100		dB
Is	Supply Current			7.2	11		7.2	11		7.2	11	mA

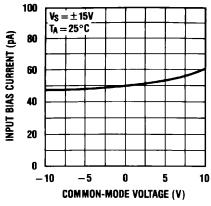
AC Electrical Characteristics (Note 7)

Symbol	Parameter	Conditions	LF147		LF347B			LF347			Units	
			Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	
	Amplifier to Amplifier Coupling	T _A =25°C,		-120			-120			-120		dB
		f=1 Hz-20 kHz										
		(Input Referred)										
SR	Slew Rate	V _S =±15V, T _A =25°C	8	13		8	13		8	13		V/µs
GBW	Gain-Bandwidth Product	V _S =±15V, T _A =25°C	2.2	4		2.2	4		2.2	4		MHz
e _n	Equivalent Input Noise Voltage	$T_A=25$ °C, $R_S=100\Omega$,		20			20			20		nV/√ Hz
		f=1000 Hz										
i _n	Equivalent Input Noise Current	T _j =25°C, f=1000 Hz		0.01			0.01			0.01		pA/√ Hz
THD	Total Harmonic Distortion	A _V =+10, R _L =10k,		<0.02			<0.02			<0.02		%
		V _O =20 Vp-p, BW=20 Hz-20 kHz										

- Note 2: Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is functional, but do not guarantee specific performance limits.
- Note 3: Unless otherwise specified the absolute maximum negative input voltage is equal to the negative power supply voltage.
- Note 4: Any of the amplifier outputs can be shorted to ground indefinitely, however, more than one should not be simultaneously shorted as the maximum junction temperature will be exceeded.
- Note 5: For operating at elevated temperature, these devices must be derated based on a thermal resistance of θ_{jA} .
- Note 6: The LF147 is available in the military temperature range $-55^{\circ}C \le T_{A} \le 125^{\circ}C$, while the LF347B and the LF347 are available in the commercial temperature range $0^{\circ}C \le T_{A} \le 70^{\circ}C$. Junction temperature can rise to T_{ij} max = 150°C.
- Note 7: Unless otherwise specified the specifications apply over the full temperature range and for $V_S=\pm20V$ for the LF147 and for $V_S=\pm15V$ for the LF347B/LF347. V_{OS} , I_B , and I_{OS} are measured at $V_{CM}=0$.
- Note 8: The input bias currents are junction leakage currents which approximately double for every 10°C increase in the junction temperature, T_j . Due to limited production test time, the input bias currents measured are correlated to junction temperature. In normal operation the junction temperature rises above the ambient temperature as a result of internal power dissipation, P_D . $T_j = T_A + \theta_{jA}$ P_D where θ_{jA} is the thermal resistance from junction to ambient. Use of a heat sink is recommended if input bias current is to be kept to a minimum.
- Note 9: Supply voltage rejection ratio is measured for both supply magnitudes increasing or decreasing simultaneously in accordance with common practice from $V_S = \pm 5V$ to $\pm 15V$ for the LF347 and LF347B and from $V_S = \pm 20V$ to $\pm 5V$ for the LF147.
- Note 10: Refer to RETS147X for LF147D and LF147J military specifications.
- Note 11: Max. Power Dissipation is defined by the package characteristics. Operating the part near the Max. Power Dissipation may cause the part to operate outside guaranteed limits.
- Note 12: Human body model, 1.5 k Ω in series with 100 pF.

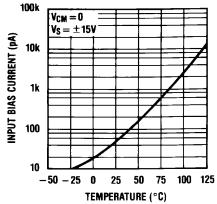
Typical Performance Characteristics





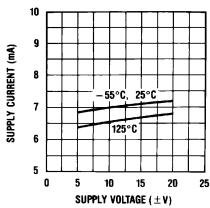
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Input Bias Current



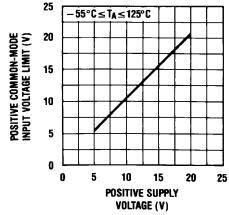
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Supply Current



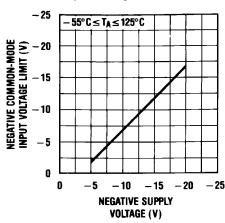
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Positive Common-Mode Input Voltage Limit



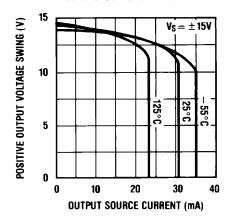
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Negative Common-Mode Input Voltage Limit

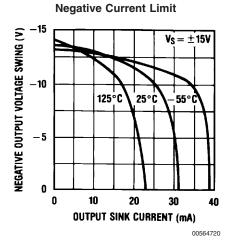


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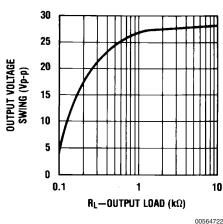
Positive Current Limit



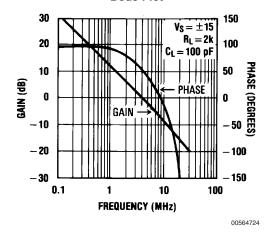
Typical Performance Characteristics (Continued)



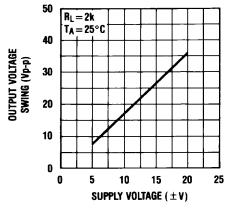




Bode Plot

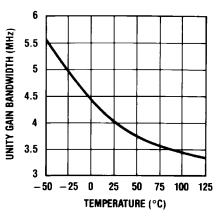


Output Voltage Swing



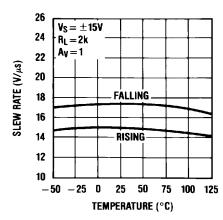
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Gain Bandwidth

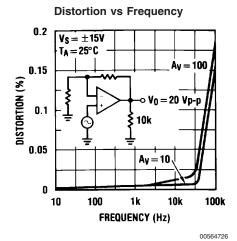


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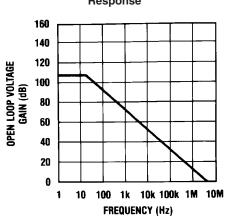
Slew Rate



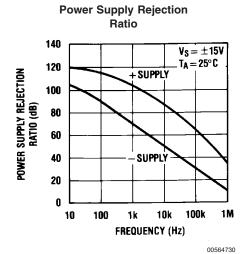
Typical Performance Characteristics (Continued)



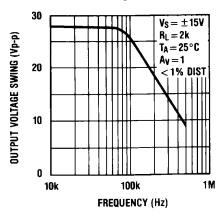
Open Loop Frequency Response



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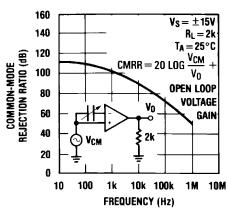


Undistorted Output Voltage Swing



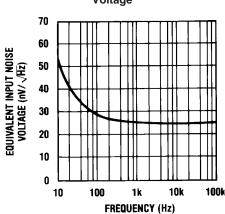
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Common-Mode Rejection



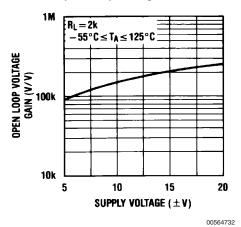
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Equivalent Input Noise Voltage

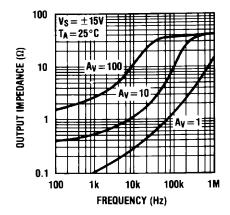


Typical Performance Characteristics (Continued)

Open Loop Voltage Gain

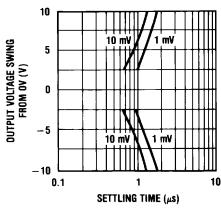


Output Impedance



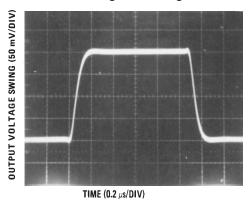
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Inverter Settling Time

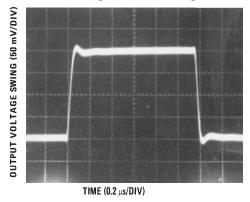


Pulse Response $R_L=2 \text{ k}\Omega, C_L=10 \text{ pF}$

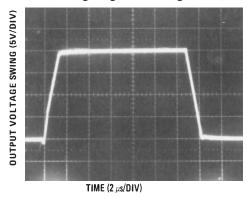
Small Signal Inverting



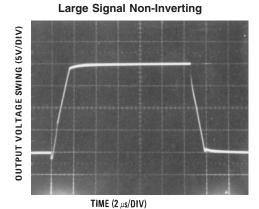
Small Signal Non-Inverting



Large Signal Inverting

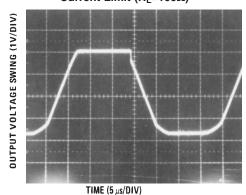


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Current Limit (R_L=100Ω)



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Application Hints

The LF147 is an op amp with an internally trimmed input offset voltage and JFET input devices (BI-FET II). These JFETs have large reverse breakdown voltages from gate to source and drain eliminating the need for clamps across the inputs. Therefore, large differential input voltages can easily be accommodated without a large increase in input current. The maximum differential input voltage is independent of the supply voltages. However, neither of the input voltages

should be allowed to exceed the negative supply as this will cause large currents to flow which can result in a destroyed unit.

Exceeding the negative common-mode limit on either input will force the output to a high state, potentially causing a reversal of phase to the output. Exceeding the negative common-mode limit on both inputs will force the amplifier output to a high state. In neither case does a latch occur since raising the input back within the common-mode range again puts the input stage and thus the amplifier in a normal operating mode.

Application Hints (Continued)

Exceeding the positive common-mode limit on a single input will not change the phase of the output; however, if both inputs exceed the limit, the output of the amplifier will be forced to a high state.

The amplifiers will operate with a common-mode input voltage equal to the positive supply; however, the gain bandwidth and slew rate may be decreased in this condition. When the negative common-mode voltage swings to within 3V of the negative supply, an increase in input offset voltage may occur.

Each amplifier is individually biased by a zener reference which allows normal circuit operation on $\pm 4.5 \text{V}$ power supplies. Supply voltages less than these may result in lower gain bandwidth and slew rate.

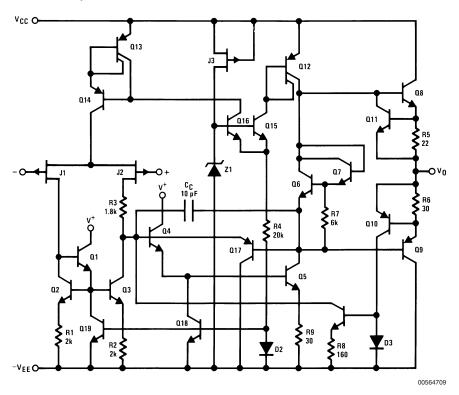
The LF147 will drive a 2 k Ω load resistance to $\pm 10V$ over the full temperature range. If the amplifier is forced to drive heavier load currents, however, an increase in input offset voltage may occur on the negative voltage swing and finally reach an active current limit on both positive and negative swings.

Precautions should be taken to ensure that the power supply for the integrated circuit never becomes reversed in polarity or that the unit is not inadvertently installed backwards in a socket as an unlimited current surge through the resulting forward diode within the IC could cause fusing of the internal conductors and result in a destroyed unit.

As with most amplifiers, care should be taken with lead dress, component placement and supply decoupling in order to ensure stability. For example, resistors from the output to an input should be placed with the body close to the input to minimize "pick-up" and maximize the frequency of the feedback pole by minimizing the capacitance from the input to ground.

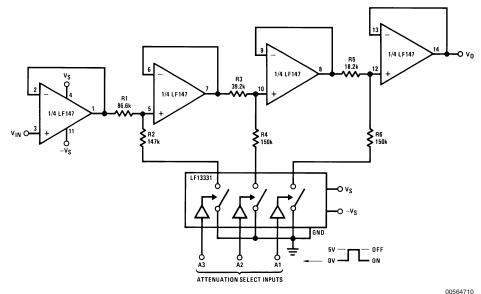
A feedback pole is created when the feedback around any amplifier is resistive. The parallel resistance and capacitance from the input of the device (usually the inverting input) to AC ground set the frequency of the pole. In many instances the frequency of this pole is much greater than the expected 3 dB frequency of the closed loop gain and consequently there is negligible effect on stability margin. However, if the feedback pole is less than approximately 6 times the expected 3 dB frequency a lead capacitor should be placed from the output to the input of the op amp. The value of the added capacitor should be such that the RC time constant of this capacitor and the resistance it parallels is greater than or equal to the original feedback pole time constant.

Detailed Schematic



Typical Applications

Digitally Selectable Precision Attenuator



All resistors 1% tolerance

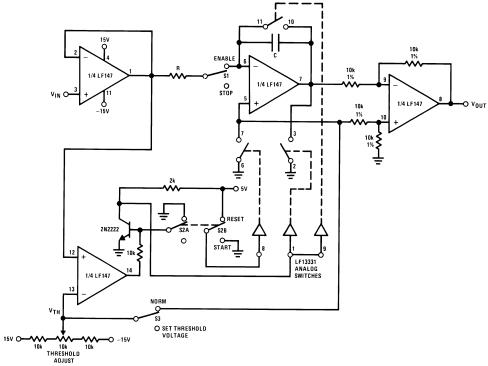
Accuracy of better than 0.4% with standard 1% value resistors
 No offset adjustment necessary

- Expandable to any number of stages
- Very high input impedance

A 1	A2	А3	Vo
			Attenuation
0	0	0	0
0	0	1	–1 dB
0	1	0	–2 dB
0	1	1	–3 dB
1	0	0	-4 dB
1	0	1	−5 dB
1	1	0	−6 dB
1	1	1	–7 dB

Typical Applications (Continued)

Long Time Integrator with Reset, Hold and Starting Threshold Adjustment



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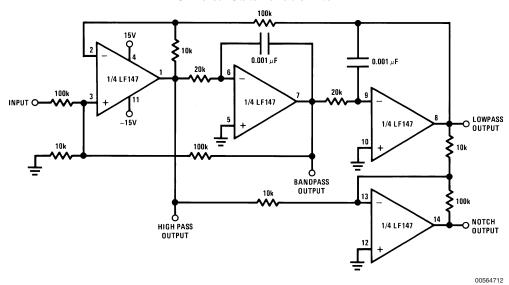
 \bullet V_{OUT} starts from zero and is equal to the integral of the input voltage with respect to the threshold voltage:

$$V_{OUT} = \frac{1}{RC} \int_0^t (V_{IN} - V_{TH}) dt$$

- Output starts when V_{IN}≥V_{TH}
- Switch S1 permits stopping and holding any output value
- Switch S2 resets system to zero

Typical Applications (Continued)

Universal State Variable Filter



For circuit shown:

 $f_0=3$ kHz, $f_{NOTCH}=9.5$ kHz

 $\Omega = 3.4$

Passband gain:

Highpass — 0.1

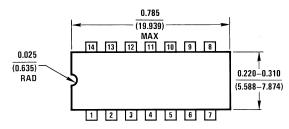
Bandpass — 1

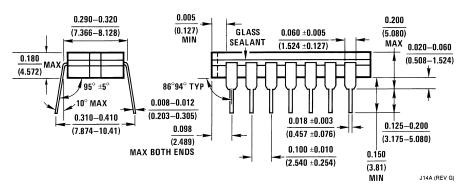
Lowpass — 1

Notch — 10

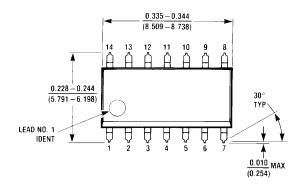
- f_oxQ≤200 kHz
- 10V peak sinusoidal output swing without slew limiting to 200 kHz
- See LM148 data sheet for design equations

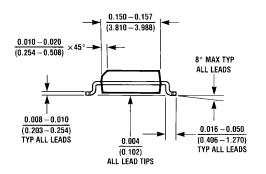
Physical Dimensions inches (millimeters) unless otherwise noted

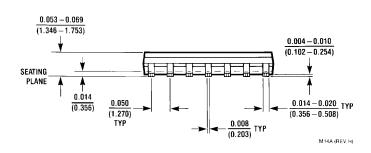




Ceramic Dual-In-Line Package (J)
Order Number LF147J, LM147J-SMD or LF147J/883
NS Package Number J14A

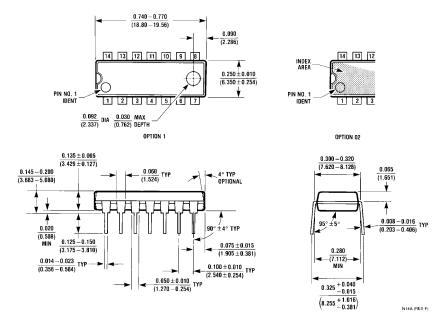






S.O. Package (M)
Order Number LF347M or LF347MX
NS Package Number M14A

Physical Dimensions inches (millimeters) unless otherwise noted (Continued)



Molded Dual-In-Line Package (N) Order Number LF347BN or LF347N NS Package Number N14A

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