



VA706
HIGH-SPEED, FAST-SETTLING
PRECISION OPERATIONAL
AMPLIFIER

T-79-07-10

FEATURES

- Fast Settling Time: $\pm 0.1\%$ in 200ns
- High Slew Rate: 42V/ μ s
- Wide Gain Bandwidth: 25MHz
- Ease of Use: Internally Compensated, Unity Gain Stable at $C_L = 50$ pF
- Large Output Current: ± 50 mA
- Low Supply Voltage Operation: ± 4 V
- Wide Input Voltage Range: Within 1.5V of V_+ and 0.5V of V_-
- Short Circuit Protection

DESCRIPTION

The VA706 is a high-speed general purpose monolithic operational amplifier useful for signal frequencies extending into the video range. The same processing innovations which permit the high speed also allow very high output currents capable of driving large capacitive loads at high speeds.

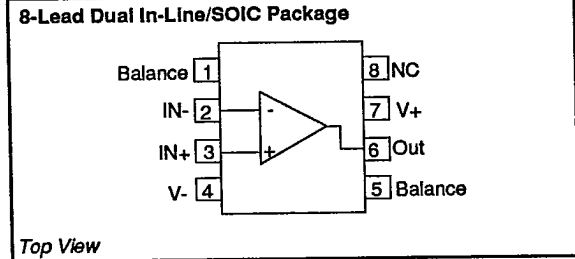
The high open-loop voltage gain of 5000V/V and high slew rate of 40V/ μ s make the VA706 ideal for analog amplification and processing of high-speed signals.

The VA706 is internally compensated for stable operation when driving capacitive loads up to 500pF. The wide gain bandwidth of 25MHz and 40V/ μ s slew rate results in $\pm 0.1\%$ settling times of 200ns, which makes the amplifier ideal for fast data conversion systems.

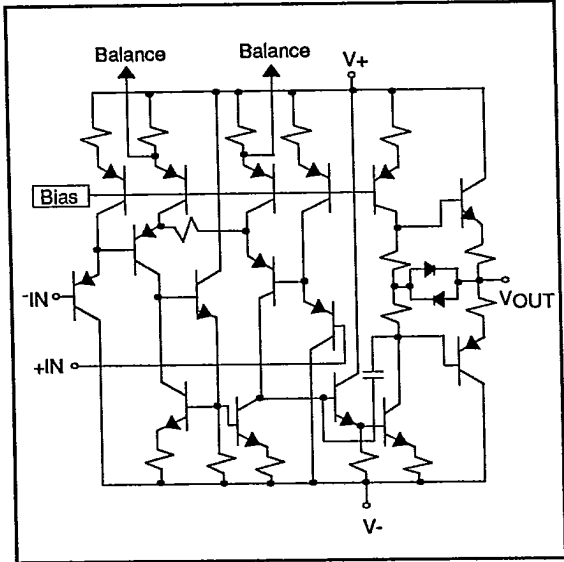
The high output current capability of ± 50 mA allows the amplifier to drive terminated transmission lines of 50 Ω with amplitudes of 5V peak to peak.

Along with the high speed and output drive capability, a 25nA offset current and trimmable offset voltage make the VA706 usable for signal conditioning applications where accuracy must be maintained.

CONNECTION DIAGRAM



SIMPLIFIED SCHEMATIC



ABSOLUTE MAXIMUM RATINGS

| | |
|--|-----------------|
| Supply Voltages | ± 6 V |
| Differential Input Voltage | ± 9 V |
| Common Mode Input Voltage | $ V_S - 0.5$ V |
| Power Dissipation (Note 1) | 450mW |
| Output Short Circuit Current Duration (Note 2) | Indefinite |
| Operating Temperature Range: | |
| Commercial (706 J, K) | 0° to 70°C |
| Storage Temperature Range | -65° to +150°C |
| Lead Temperature (Soldering to 60 Sec.) | 300°C |

Note 1: Power derating above $T_A = 70^\circ\text{C}$ to be based on a maximum junction temperature of 150°C and the following thermal resistance factors:

Note 2: Continuous short circuit protection is allowed to the following case and ambient temperatures:

| PKGE. | θ_{JC} (°C/W) | θ_{JA} (°C/W) | T_C (°C) | T_A (°C) |
|-------|----------------------|----------------------|------------|------------|
| DIP | 75 | 180 | 110 | 70 |
| SOIC | 115 | 180 | 95 | 70 |

PACKAGE TYPES AVAILABLE

- 8-Pin Plastic DIP
- 8-Pin Cerdip
- 8-Pin SOIC

T-79-07-10

ELECTRICAL CHARACTERISTICS ($V_S = \pm 5V$, $T_A = 25^\circ C$ unless otherwise stated)

| PARAMETER | SYMBOL | CONDITIONS | VA706J | | | VA706K | | | UNITS |
|--|--------------------------|---|-----------|--------------|------|-----------|--------------|------|------------------|
| | | | MIN | TYP | MAX | MIN | TYP | MAX | |
| Input Offset Voltage T_{Min} to T_{Max} | V_{OS} | | | 8 | 20 | | 4 | 10 | mV |
| | | $0^\circ \leq T_A \leq 70^\circ C$ | | 11 | 28 | | 6 | 16 | |
| Average Offset Voltage Drift | $\Delta V_{OS}/\Delta T$ | $0^\circ \leq T_A \leq 70^\circ C$ | | 20 | | | 20 | | $\mu V/^\circ C$ |
| Input Bias Current | I_B | | | 650 | 1100 | | 650 | 1100 | nA |
| Input Offset Current T_{Min} to T_{Max} | I_{OS} | | | 35 | 120 | | 35 | 120 | nA |
| | | $0^\circ \leq T_A \leq 70^\circ C$ | | 70 | 200 | | 70 | 200 | |
| Input Common Mode Range | V_{CM} | | +3 -4 | +3.5 -4.5 | | +3 -4 | +3.5 -4.5 | | V |
| Differential Input Resistance | R_{IND} | (Note 1) | 3 | 10 | | 3 | 10 | | $M\Omega$ |
| Common Mode Input Resistance | R_{INC} | (Note 1) | 4 | 8 | | 4 | 8 | | $M\Omega$ |
| Differential Input Capacitance | C_{IND} | (Note 1) | | 2 | | | | | pF |
| Common Mode Input Capacitance | C_{INC} | (Note 1) | | 3 | | | 3 | | pF |
| Input Voltage Noise | e_N | BW=10Hz to 100KHz | | 12 | | | 12 | | $\mu VRMS$ |
| Open Loop Voltage Gain | A_V | $V_{OUT} = \pm 3V$ $R_L = 2k\Omega$ | 1 | 5 | | 2 | 5 | | V/mV |
| Output Voltage Swing | V_{OUT} | $R_L = 2k\Omega$ | ± 3.5 | +4 -4.2 | | ± 3.5 | +4 -4.2 | | V |
| | | $R_L = 51\Omega$ | ± 2.0 | ± 2.4 | | ± 2.5 | ± 2.7 | | |
| Power Supply Current | I_S | | | 7 | 10 | | 7 | 10 | mA |
| Common Mode Rejection Ratio | CMRR | $V_{CM} = \pm 2V$ | 60 | 70 | | 60 | 70 | | dB |
| Power Supply Rejection Ratio | PSRR | $\Delta V_{PS} = \pm 0.5V$ | 60 | 66 | | 60 | 66 | | dB |
| Slew Rate | SR | 10-90% of Leading Edge (Figure 1) | 30 | 42 | | 30 | 42 | | V/ μs |
| Settling Time | t_S | To $\pm 0.1\%$ ($\pm 4mV$) of Final Value (Figure 1) (Note 1) | | 200 | 250 | | 200 | 250 | ns |
| Gain Bandwidth Product | GBW | | | 25 | | | 25 | | MHz |
| Small Signal Rise/Fall Time | t_r / t_f | $e_O = \pm 50mV$ 10-90% (Figure 1) | | 7 | | | 7 | | ns |
| Full Power Bandwidth | BW_{FP} | $R_L = 2k\Omega$ $C_L = 50pF$ $V_{OUT} = 6Vp-p$ | | 2.2 | | | 2.2 | | MHz |

Notes: 1. Not tested, guaranteed by design.

T-79-07-10

DIE INFORMATION

| WAFER TEST LIMITS | | | | |
|--|-----------|--|------------------------|----------------|
| $V_S = \pm 5V, T_A = 25^\circ C$ unless otherwise stated | | | | |
| PARAMETER | SYM | CONDITIONS | VA706XS LIMIT | UNITS |
| Input Offset Voltage | V_{OS} | | 20 | mV Max |
| Input Bias Current | I_B | | 1000 | nA Max |
| Input Offset Current | I_{OS} | | 50 | nA Max |
| Input Common Mode Range | V_{CM} | | +3 -4 | V Min |
| Open Loop Voltage Gain | A_V | $V_{OUT} = \pm 3V$ $R_L = 2k$ | 2 | V/mV Min |
| Output Voltage Swing | V_{OUT} | $R_L = 2k \Omega$ $R_L = 51 \Omega$ | ± 3.5 ± 2.5 | V Min |
| Power Supply Current | I_S | | 10 | mA Max |
| Common Mode Rejection Ratio | CMRR | $V_{CM} = \pm 2V$ | 60 | dB Min |
| Power Supply Rejection Ratio | PSRR | $\Delta V_{PS} = \pm 0.5V$ | 60 | dB Min |
| Slew Rate | SR | 10-90% of Leading Edge (Figure 1a,b) | 30 | V/ μs Min |

| TYPICAL ELECTRICAL CHARACTERISTICS | | | | |
|--|-------------|---|-----------------|-------|
| $V_S = \pm 5V, T_A = 25^\circ C$ unless otherwise stated | | | | |
| PARAMETER | SYM | CONDITIONS | VA706XS TYPICAL | UNITS |
| Input Offset Voltage | V_{OS} | | 30 | mV |
| Input Offset Current | I_{OS} | | 75 | nA |
| Settling Time | t_S | To $\pm 0.1\%$ of Final Value (Figure 1a,b) | 200 | ns |
| Gain Bandwidth Product | GBW | | 25 | MHz |
| Small Signal Rise/Fall Time | t_r / t_f | $e_O = \pm 50mV$ 10-90% (Figure 1c) | 7 | ns |
| Full Power Bandwidth | BW_{FP} | $R_L = 2k \Omega$ $C_L = 50pF$ $V_{OUT} = 6V_{p-p}$ | 2.2 | MHz |

DICE POLICY

Electrical Characteristics

Each die is electrically tested to the commercial or military grade DC parameters to guard band limits at 25°C to guarantee operation over the full temperature range.

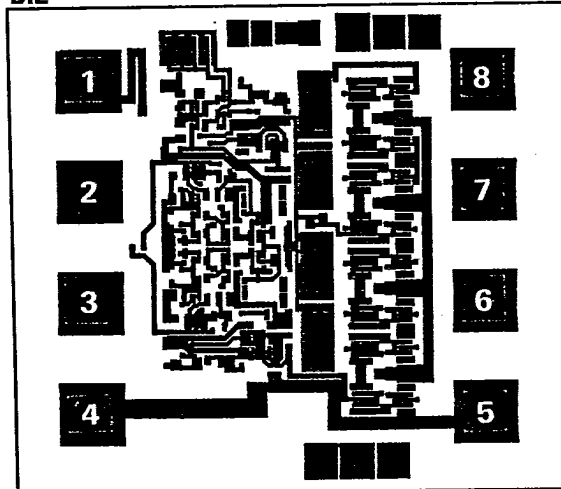
Quality Assurance

All dice are 100% visually inspected to the requirement of MIL-STD-883C, Method 2010.2, Condition 3.
All dice are glass passivated with only the bonding pads exposed to provide scratch protection.
All dice are provided with gold backing.

Shipping Packages/Order Information

All dice are packaged in die crates with individual compartments which prevent damage to the die during shipping. The individual cavity size of the die crate is such that maximum rotation of the die within the cavity is < 45°.
Minimum order for dice is 100, supplied only in multiples of 100.

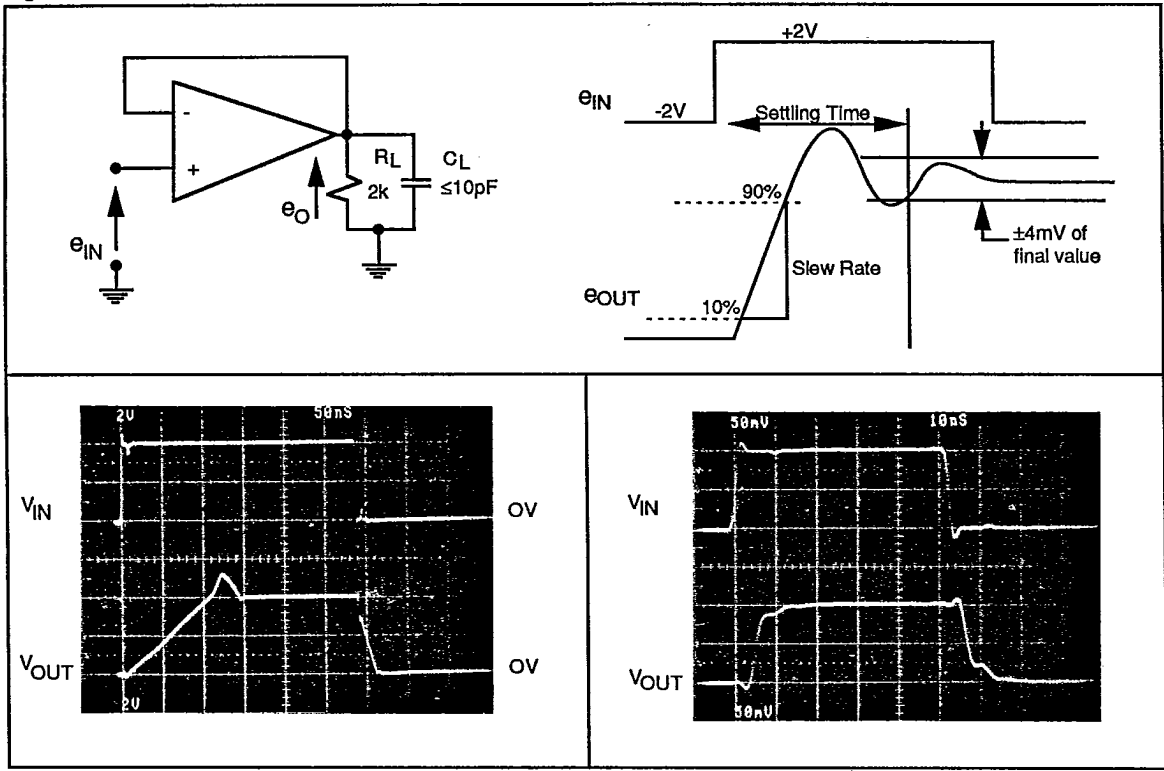
DIE



Die size = 0.035 x 0.035 inch (1225 sq. mils)
0.89 x 0.89 mm (0.79 sq. mm)
Shipped in die crates.

T-79-07-10

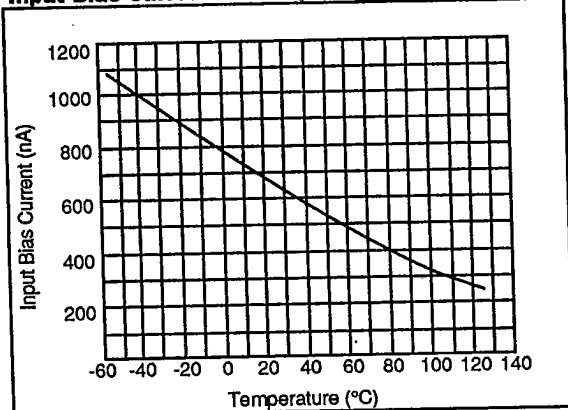
Figure 1: Slew Rate and Settling Time Test Circuit



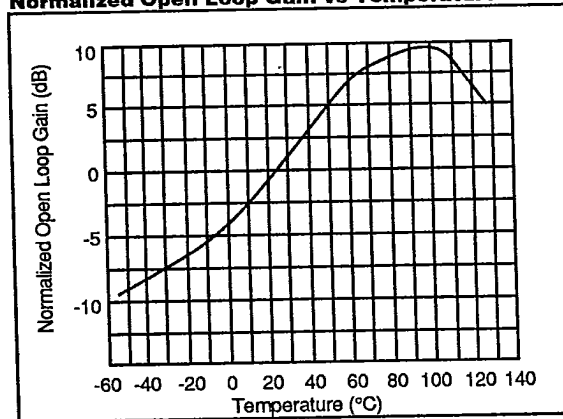
T-79-07-10

TYPICAL PERFORMANCE CHARACTERISTICS ($V_S = \pm 5V$, $T_A = 25^\circ C$ unless otherwise stated)

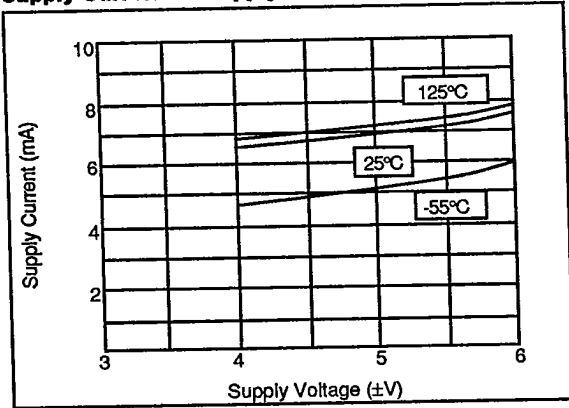
Input Bias Current vs Temperature



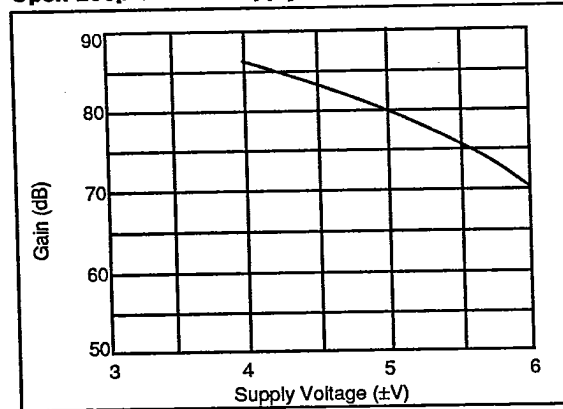
Normalized Open Loop Gain vs Temperature



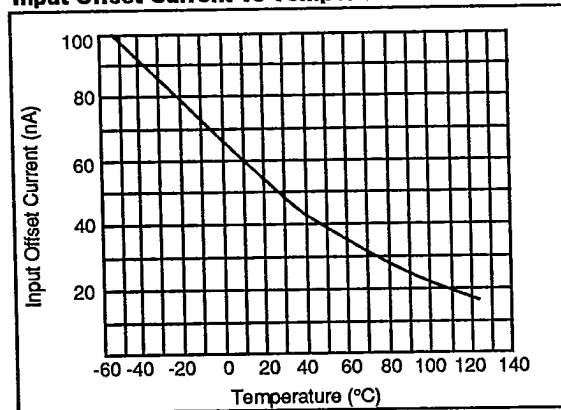
Supply Current vs Supply Voltage



Open Loop Gain vs Supply Voltage



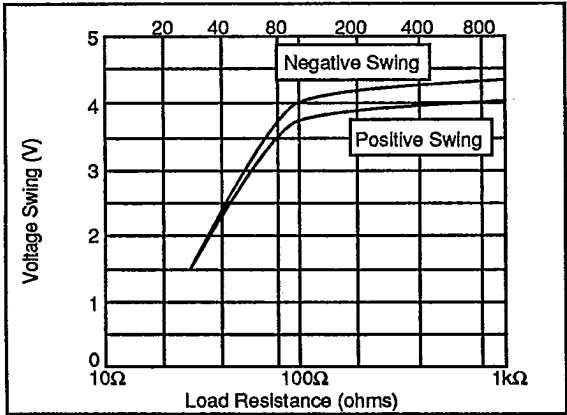
Input Offset Current vs Temperature



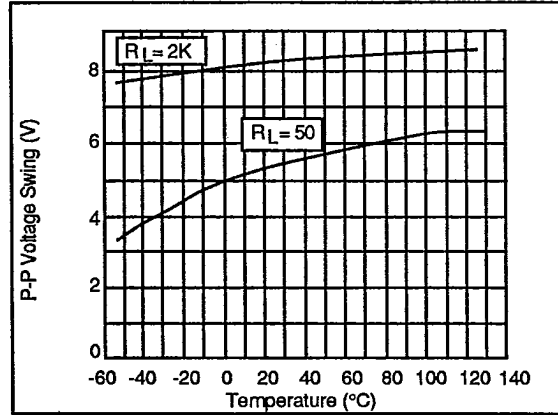
T-79-07-10

TYPICAL PERFORMANCE CHARACTERISTICS ($V_S = \pm 5V$, $T_A = 25^\circ C$ unless otherwise stated)

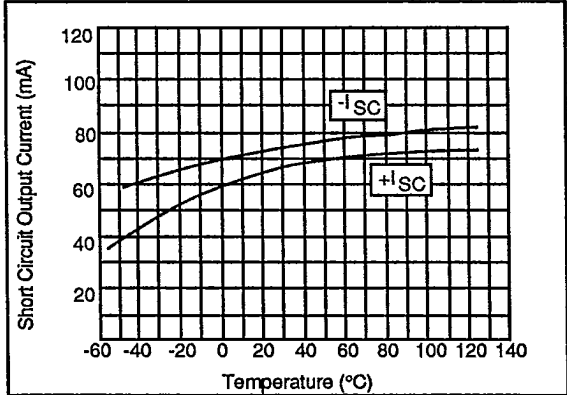
Maximum Output Voltage Swing vs Load Resistance



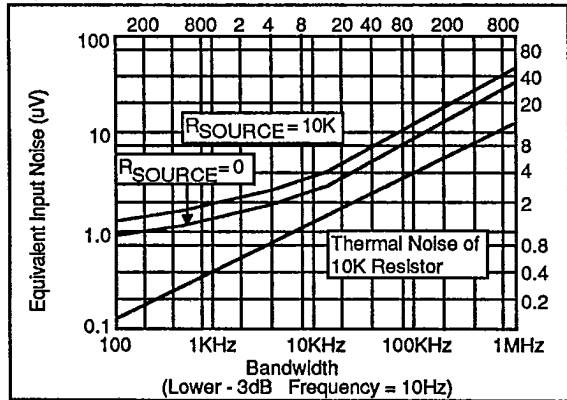
Maximum Output Voltage Swing vs Temperature



Short Circuit Output Current vs Temperature



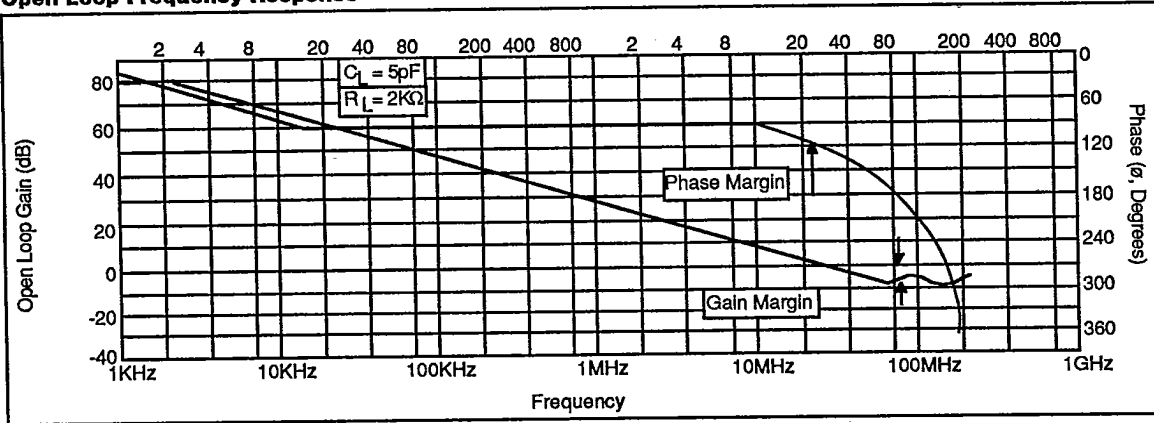
Equivalent Input Noise vs Bandwidth



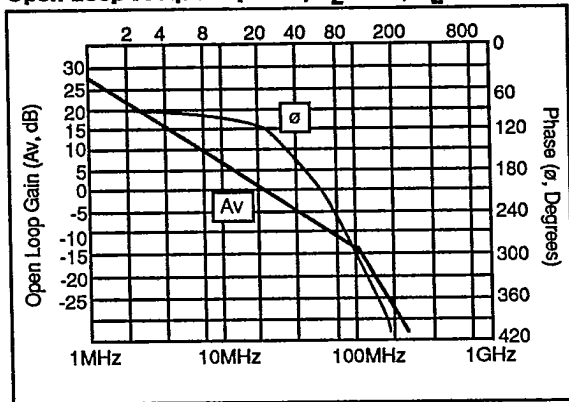
T-79-07-10

TYPICAL PERFORMANCE CHARACTERISTICS ($V_S = \pm 5V$, $T_A = 25^\circ C$ unless otherwise stated)

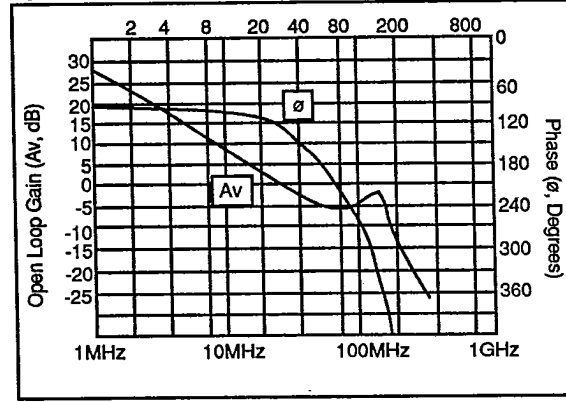
Open Loop Frequency Response



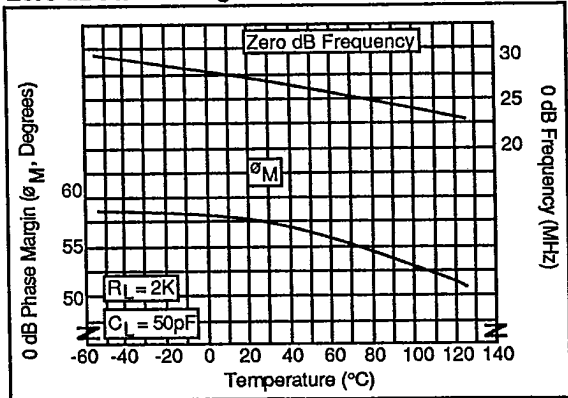
Open Loop Freq. Response, $R_L = 50\Omega$, $C_L = 50pF$



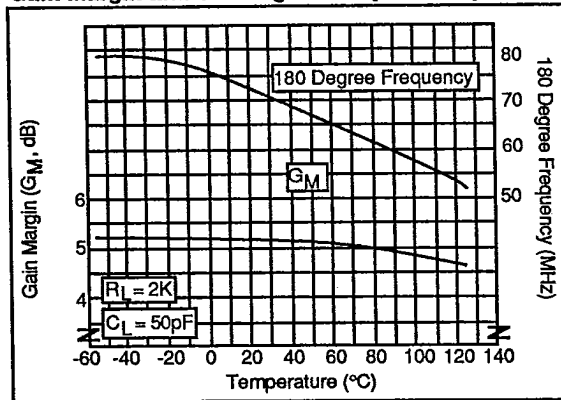
Open Loop Freq. Response, $R_L = 2K\Omega$, $C_L = 50pF$



Zero dB Phase Margin and Zero dB Freq. vs Temp.



Gain Margin and 180 Degree Freq. vs Temp.



T-79-07-10

APPLICATION INFORMATION

AC Characteristics

The 28MHz 0dB crossover point of the VA706 is achieved without feed-forward compensation, a technique which can produce long tails in the recovery characteristic. The single pole rolloff follows the classic 20dB/decade slope to frequencies approaching 50MHz. The phase margin of 58°, even with a capacitive load of 50pF, gives stable and predictable performance down to unity gain follower configurations.

At frequencies beyond 50MHz, the 20dB/decade slope is disturbed by an output stage zero, the damping factor of which is dependent upon the load capacitor. This results in loss of gain margin (gain at loop phase = 360°) at frequencies of 70 to 100MHz which at a gain margin of 5dB ($R_L = 2k$, $C_L = 50pF$) results in a 10dB peak in the unity gain follower closed loop characteristic (Figure 2).

Figure 2 shows a blow up of the open loop characteristics in the 10MHz to 200MHz frequency range as well as the corresponding unity gain follower characteristics at similar load conditions. It is seen that the output stage zero results in bandwidth extension beyond the 28MHz, 0dB crossover point. In fact, with the proper choice of the R_L, C_L load, the unity gain follower can be "tweaked" to give flat small signal response to 100MHz.

Figure 3 shows corresponding time domain response for a small signal step. As expected there is a strong 80MHz ring for $R_L = 2k\Omega$, $C_L = 50pF$ which disappears at $R_L = 50\Omega$, $C_L = 5pF$.

Offset Voltage Nulling

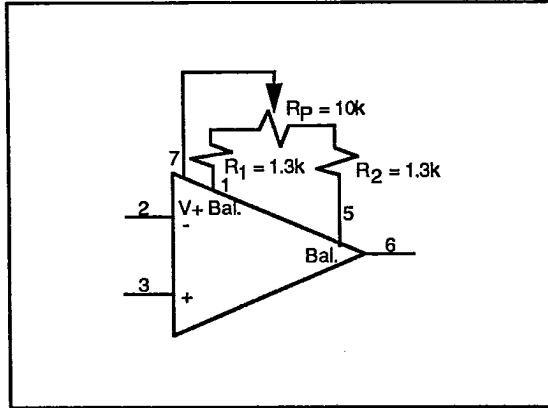
The configuration of Figure 4 will give a typical V_{OS} nulling range of $\pm 15mV$. If a smaller adjustment range is desired, resistor values $R_1 = R_2$ can be increased accordingly. For example, at $R_1 = 3.6k\Omega$, the adjustment range is $\pm 5mV$. Since pins 1 and 5 are not part of the signal path, AC characteristics are left undisturbed.

Layout Considerations

As with any high-speed wideband amplifier, certain layout considerations are necessary to ensure stable operation. All connections to the amplifier should be kept as short as possible, and the power supplies bypassed with 0.1 μF capacitors to signal ground. It is suggested that a ground plane be considered as the best method for ensuring stability because it minimizes stray inductance and unwanted coupling in the ground signal paths.

To minimize capacitive effects, resistor values should be kept as small as possible, consistent with the application.

Figure 4: Vos Nulling Method



T-79-07-10

Figure 2: Unity Gain Follower Frequency Characteristics

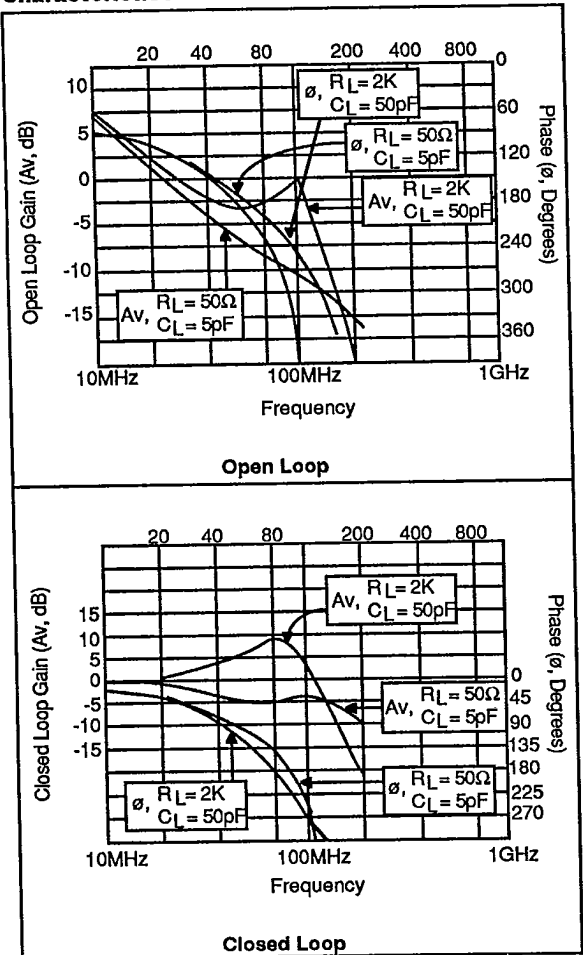


Figure 3: Unity Gain Follower Step Response

